



















Kalaburagi - 585103, Karnataka - India ಕಲಬುರಗಿ 585 103 ಕರ್ನಾಟಕ - ಬಾರತ

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A REPORT ON INTRODUCTION TO CIRCUIT AND LAYOUT DESIGN

(A three months value added course for Third-Year UG Program)

Academic year 2022-2023



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FACULTY OF ENGINEERING AND TECHNOLOGY [CO-EDUCATION] DEPARTMENT OF ELECTRONICS AND COMMUNICATION ENGINEEERING VALUE ADDED COURSE

ON

"Introduction to Circuit and layout Design"

About the Course: The value added course on "Introduction to Circuit and layout Design", aim to introduce the most demand skills in VLSI. This course deals with the fundamental theory to Practical hand on training in VLSI. This course allows the student to learn from the scratch to write code, design using the CMOS transistors which help them to manage their projects efficiently and make the career as a VLSI design engineer.

High light of the Course:

- 30 hours Course Duration
- Introduction to Verilog & CMOS
- Practical training: How to write code using Verilog HDL Programming.
- Practical training: How to design Semicustom VLSI circuits using Cadence tools.
- Mini Project based on 45nm node
- Certificate

(Min of 75% attendance)

Target Audience: UG / PG

Course Experts: Dr. Anuradha Savadi, Associate Professor, Faculty of Engineering and Technology [coeducation], Sharnbasva University, Kalaburagi.

Alumni: Alumni Industry Engineers

Advisory Committee:

- Dr. Shivkumar Jawaligi, Professor & Dean, Faculty of Engineering and Technology [CO-EDUCATION],
 Sharnbasva University, Kalaburagi.
- Dr. Shashidhar Sonnad, Professor & Chairman, Faculty of Engineering and Technology [CO-EDUCATION], Sharnbasva University, Kalaburagi.

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Subject Code

Number Lecture Hour/Week

Hands on tool access

date. 21 July 2005

Text Book:

Design any combinational and sequential circuits.

Draw the layout for various schematics and run DRC,LVS

1. Behzad Razavi, —Design of Analog CMOS Integrated Circuits, TMH, 2007.

2. Alan Hastings- Art of Analog Layout, Edition. 2nd · Publisher. Pearson · Publication

Number of Lecture Hours









CIE Marks

SEE Marks

Exam Hours

0

0

0



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INTRODUCTION TO CIRCUIT AND LAYOUT DESIGN
[As per Choice Based Credit System (CBCS) scheme]

SEMESTER – VI

CREDITS-00

VALCLD-1

04

30

Course Objectives: This course will enable students to:

Course Objectives. This course will chaose students to.	
 Learn to work in a Linux environment 	
 Understand the complete physical design flow 	
Plan and execute implementations of systems	
 Determine the performance of a system through simulation 	
Establish capability for CAD tool development and enhancement	
Modules	Teaching
	Hours
Module -1	110415
Basic electronics: BJT,MOSFET, diode, capacitor, resistor, inductor, current,	6 Hours
voltage, ohm's law, KCL & KVL, conductor, insulator, semiconductor, STI	
(static time analysis) lab- schematic for NAND3, NOR3, AND4, OR4.	
Module -2	1
CMOS Technology: Fabrication ,CMOS inverter characteristics, lab-logic gate	6 Hours
design, single stage amplifier (cadence 45nm tool)	
Module -3	1
Introduction to Digital layout: Standard cells, layout introduction, stick	6 Hours
diagrams lab- standard cell design of basic gates.	
Module -4	
Introduction to Analog layout: process variations, layout dependent effects.	6 Hours
Lab- any amplifier design.	
Module -5	
Physical design verification: LVS, DRC, DFM, HVDRC, Antenna effects	6 Hours
Course outcomes: After studying this course, students will be able to:	1
Understand device physics and technology	
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Estd. 2017 ನ್ಯಾಪನೆ : 2017

















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Reference Books:

- **1.** Sung Mo Kang &YosufLeblebici, "CMOS Digital Integrated Circuits: Analysis and Design", Tata McGraw-Hill, Third Edition.
- 2. Neil Weste and K. Eshragian, "Principles of CMOS VLSI Design: A System Perspective", Second Edition, Pearson Education (Asia) Pvt. Ltd. 2000.

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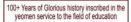
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BACKEND VALUE ADDED COURSE

DEPT OF ECE, FACULTY OF ENGG AND TECHNOLOGY [CO-EDUCATION] SUBJECT: CIRCUIT AND LAYOUT DESIGN

SI NO	STUDENT NAME	USN
1	AKANSHA	SG20ECE015
2	AADIL KHAN	SG20ECE013
3	ABDUL RAHMAN	SG20ECE001
4	ABHISHEK ALLAGI	SG20ECE002 SG20ECE005
5		SG20ECE003 SG21ECE501
	ANAND ANAND MANAJI	
6		SG20ECE017
7	ARCHANA	SG20ECE020
8	ARUN	SG20ECE021
9	BHARGAVI	SG20ECE025
10	BHOOMIKA DESAI	SG20ECE028
11	DEEPAK	SG20ECE032
12	GEETA	SG20ECE039
13	HARSHVARDHAN GAIKWAD	SG20ECE041
14	JYOTI	SG20ECE043
15	KARAN	SG21ECE506
16	MAHESH	SG20ECE050
17	MAHESH LADVANTHI	SG20ECE051
18	MALLIKARJUN	SG20ECE052
19	MANIKANTH	SG20ECE507
20	MD. MUBARAK	SG20ECE057
21	NAGARAJ	SG20ECE058
22	NIRANJAN	SG20ECE062
23	PRADEEP PATIL	SG20ECE066
24	PRADEEP	SG20ECE067
25	PRADEEP N D	SG20ECE068
26	PRASHANTH	SG20ECE509
27	RADHIKA	SG20ECE071
28	RAHUL SINGH	SG20ECE072
29	S SANNIDHI	SG20ECE073
-30	SAKSHI	SG20ECE074
31	SHAHEDA	SG20ECE064
32	SHARAN KUMAR	SG20ECE076
33	SHIVANI	SG20ECE078
34	SHRADDHA R P	SG20ECE081
35	SUDEEP EKLOOR	SG20ECE089
36	SUNIL KUMAR	SG20ECE091
37	SUPRIYA	SG20ECE092



















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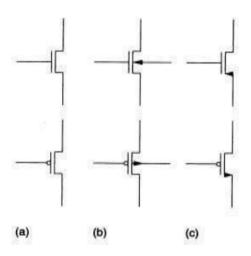
UNIT-1

MOS TRANSISTOR PRINCIPLE

INTRODUCTON

An MOS transistor is a majority-carrier device, in which the current in a conducting channel between the source and the drain is modulated by a voltage applied to the gate.

Symbols



NMOS

PMOS

NMOS (n-type MOS transistor)

- (1) Majority carrier = electrons
- (2) A positive voltage applied on the gate with respect to the substrate enhances the number of electrons in the channel and hence increases the conductivity of the channel.
- (3) If gate voltage is less than a threshold voltage Vt, the channel is cut-off (very low current between source & drain).

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PMOS (p-type MOS transistor)

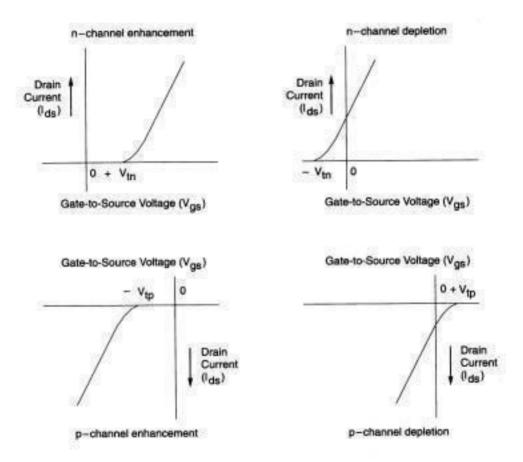
- Majority carrier = holes
- Applied voltage is negative with respect to substrate.

Threshold voltage (Vt):

- The voltage at which an MOS device begins to conduct ("turn on")
- Relationship between Vgs (gate-to-source voltage) and the source-todrain current (Ids), given a fixed drain-to-source voltage (Vds).



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- Devices that are normally cut-off with zero gate bias are classified as "enhancement-mode "devices.
- Devices that conduct with zero gate bias are called "depletionmode devices
- Enhancement-mode devices are more popular in practical use.

NMOS Enhancement Transistor

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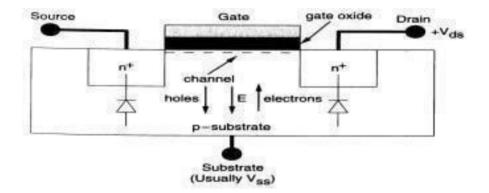


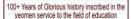




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It Consist of

- Moderately doped p-type silicon substrate
- Two heavily doped n + regions, the source and drain, are diffused
- Channel is covered by a thin insulating layer of silicon dioxide (SiO2) called
 " Gate Oxide "
- Over the oxide is a polycrystalline silicon (polysilicon) electrode, referred to as the "Gate"

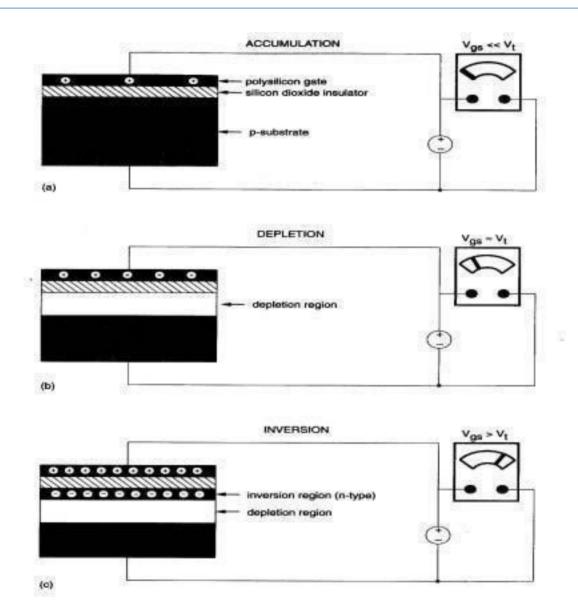
Features

- Since the oxide layer is an insulator, the DC current from the gate to channel is essentially zero.
- No physical distinction between the drain and source regions.
- Since SiO2 has low loss and high dielectric strength, the application of high gate fields is feasible.
- In operation
- \triangleright Set Vds > 0 in operation
- > Vgs =0 no current flow between source and drain. They are insulated by two reversed-biased PN junctions
- When Vg > 0, the produced E field attracts electrons toward the gate and repels holes.
- If Vg is sufficiently large, the region under the gate changes from p-type to n-type(due to accumulation of attracted elections) and provides a conducting path between source and drain.
- The thin layer of p-type silicon is said to be "inverted".
- Three modes (see Fig 2.4)
 - Accumulation mode (Vgs << Vt)
 - o Depletion mode (Vgs =Vt)
 - \circ Inversion mode (Vgs > Vt)



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Electrically

- (1) An MOS device can be considered as a voltage-controlled switch that conducts when Vgs >Vt (given Vds>0)
- (2) An MOS device can be considered as a voltage-controlled resistor

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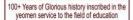
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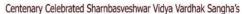
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Effective gate voltage (Vgs-Vt)

At the source end , the full gate voltage is effective in the inverting the channel.

At the drain end , only the difference between the gate and drain voltage is effective















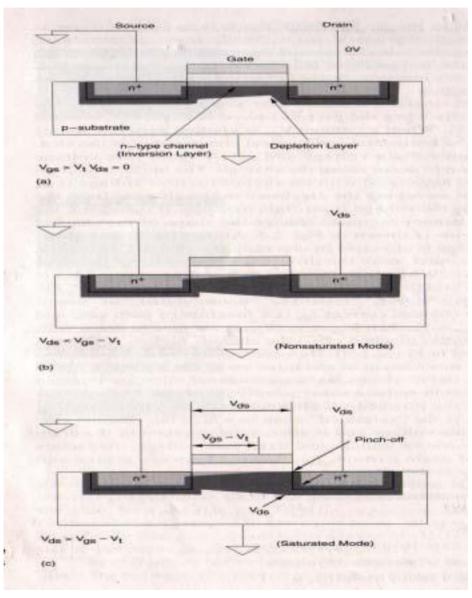






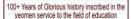


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Pinch-off

- Vds > Vgs-Vt => Vgd < Vt => Vd > Vg-Vt (Vg is not big enough)
- The channel no longer reaches the drain. (Fig 2.5 c)
- As electrons leave the drain depletion region and are subsequently accelerated toward the drain.























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The voltage across the pinched-off region remains at (Vgs-Vt) =>"saturated" state in which the channel current as controlled by Vg, and is independent of Vd

For fixed Vds and Vg, Ids is function of

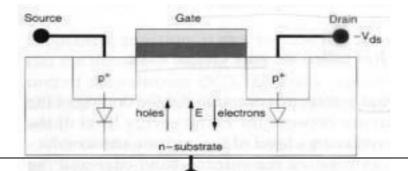
- Distance between drain & source
- Channel width
- > Vt
- > Thickness of gate oxide
- The dielectric constant of gate oxide
- (6) Carrier (hole or electron) mobility, .

Conducting mode

- "cut-off" region : Ids ≈ 0, Vgs < Vt
- "Nonsaturated" region: weak inversion region, when Ids depends on Vg & Vd
- "Saturated" region: channel is strongly inverted and Ids is ideally independent of Vds (pinch-off region)
- "Avalanche breakdown" (pinch-through): very high Vd => gate has no control over Ids

PMOS Enhancement Transistor

- Vg < 0
- Holes are major carrier
- Vd < 0, which sweeps holes from the source through the channel to the drain.



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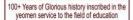


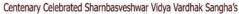
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Threshold voltage

- A function of
- Gate conductor material
- Gate insulator material
- Gate insulator thickness
- Impurity at the silicon-insulator interface
- Voltage between the source and the substrate Vsb
- Temperature
 - a. -4 mV/'C high substrate doping
 - b. -2 mV/'C low substrate doping



















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SCALING

In scaling there are really two issues

• Devices Can we build smaller devices

What will their performance be try to avoid the wet noodle effect Wires

There is concern about our ability to scale both of these Components

Limitations

Limitations to device scaling has been around since working in 3m nMOS, 22 years ago (actually bipolar)

- Worries were
 - Short channel effect
 - Punchthrough
- drain control of current rather than gate
 - Hot electrons
 - Parasitic resistances
- · Now worries are a little different
 - Oxide tunnel currents
 - Punchthrough
 - Parameter

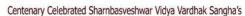
control Parasitic resistances

Transistor scaling

People are building very short channel devices

- Shown are I-V curves for 15nm L pMOS
- And a short channel nMOS
- The structure is strange
- FinFET
- But you can make them work











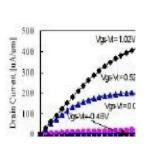


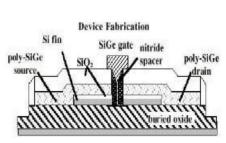


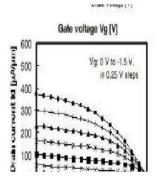




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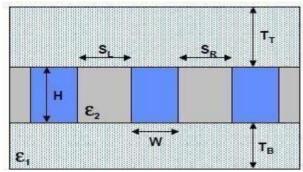




Wire scaling

More uncertainty than transistor scaling

- Many options with complex trade-offs
- For each metal layer
- Need to set H, TT, TB, e1, e2, conductivity of the metal

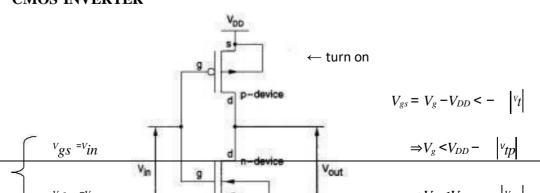


ILD_{Top}

Metal, ILD_{middle}

ILD_{Bot}

CMOS INVERTER



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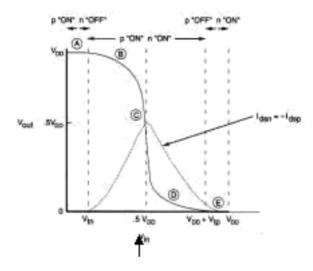






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Both transistors are "on"

$$P = fcv^2 \cdot \alpha$$

(Switching activity)

Solve for

$$\begin{cases} I_{dsn} = -I_{dsp} \\ V_{inn} = V_{inp} \end{cases}$$

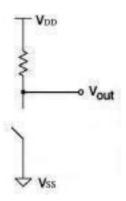
(1) Region A.

$$0 \le V_{in} \le V_{tn}$$

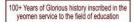
n-device is 'off',
$$I_{dsn} = 0$$
 (= $-I_{dsp}$)

p-device is in 'linear' mode

$$^{V}out^{-V}DD^{=V}ds\overline{p}^{=0}$$



$$^{V}tn \leq V_{in} \leq ^{VDD}$$





















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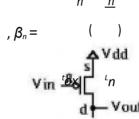
2 p-device: linear mode n-device : saturation mode $[\textit{Vi} \; - \textit{Vt} \;]^2$ <u>με</u> W n n

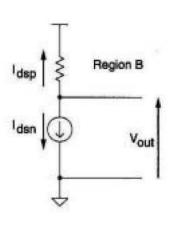
 $n: \int ds n^{-\beta} n$

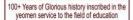
p :

∨gs ^{=v}in

 $^{V}ds = ^{V}out - V_{DD}$

























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(3) Region C. PMOS, NMOS: saturation

<u>β</u>

$$dsp = - (V_{in} - V_{DD} - V_{tp})^{2}$$

$$\frac{2}{\underline{\beta}_{n}}$$

$$\int dsn = (V_{in} - V_{tn})^2$$

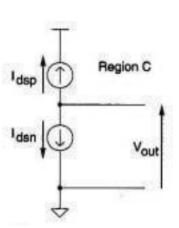
2

with
$$\int dsp^{-1} dsn$$

$$^{V}DD^{+V}tp + V_{tn}$$
 β

р

 \Rightarrow by setting $\beta_n = \beta_p$ and $V_{tn} = -V_{tp}$



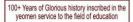
we have



: one value only

possible Vout

⊂ ^vin ^{-v}tn ^{<v}out





















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N-MOS

$$^{V}gs^{-V}tn < V_{ds}$$

P-MOS
$$\begin{cases} (V_{DD} - V_{in}) - V_{tp} > (V_{DD} - V_{out}) \\ V_{gs} - V_{tp} > V_{ds} \\ \Rightarrow^{V} out <^{V} in - ^{V} tp \end{cases}$$

$$\Rightarrow V_{in} - V_{tn} \leq V_{out} \leq V_{in} - V_{tp}$$

Negative value

 $\Rightarrow V_{in}$ is fixed at ${}^V\!DD$, V_{out} varies

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N-MOS: linear mode

$$I_{dsp} = -\frac{1}{\beta_p} (V_{in} - V_{DD} - V_{tp})^2$$

$$V \quad 2$$

$$O_t u$$

$$I_{dsn} = \beta_n [(V_{in} - V_{tn})^{V_{out}} - \frac{O_t u}{O_t}]$$

$$2$$

solve
$$\int dsn^{-1} dsp$$

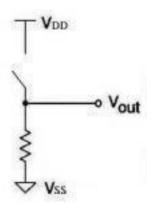
$$\Rightarrow Vou=(V -Vt) - (Vi-Vt) 2 - {}^{\beta}p_{(V}i-VD-Vt)^{2}$$

$$t \quad in \quad n \quad \sqrt{n \quad n \quad \beta \quad n \quad D \quad p}$$

(4) Region E.
$$vin \ge vDD + V_{tp}$$

$$\rightarrow p\text{-device 'off'} \qquad (n\text{-device is in 'linear' mode})$$

$$\stackrel{\rightarrow}{\rightarrow} tdsp = 0 \implies term = 0$$



see Table 2.3 for summary















Sharnbas



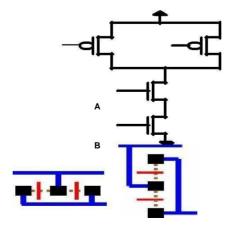
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LAYOUT DESIGN RULES

Why we need design rules

- Masks are tooling for manufacturing.
- Manufacturing processes have inherent limitations in accuracy.
- Design rules specify geometry of masks which will provide reasonable yields.
- Design rules are determined by experience.

Layers		Connection Rules						
	Metal (BLUE)				poly	n-diff	p-diff	metal
	Polysilicion (RED)	poly	S	N	Р	NC
	N-Diffusion	(Green)		n-diff		S	X	NC
	P -Diffusion	(Brown)		p-diff			S	NC
	Contact / Via			metal				S











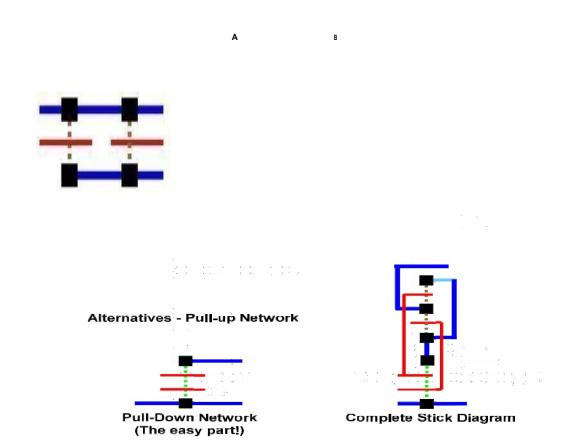




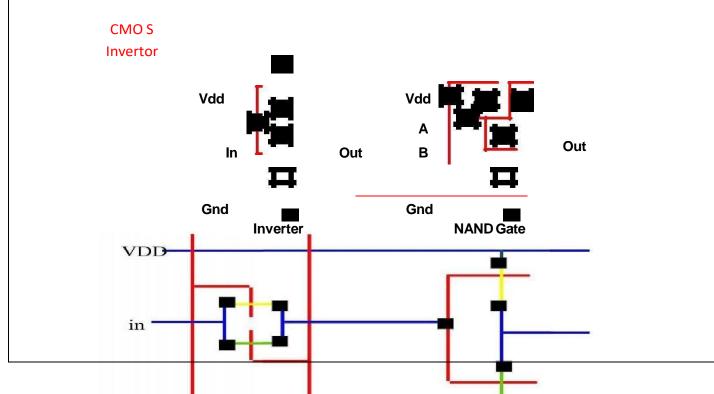


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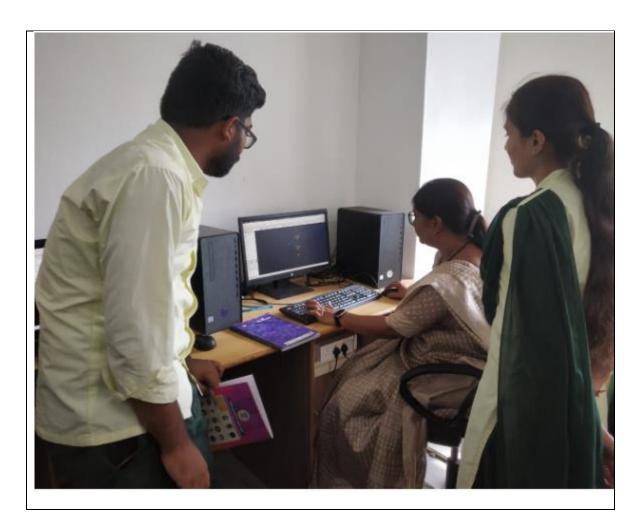








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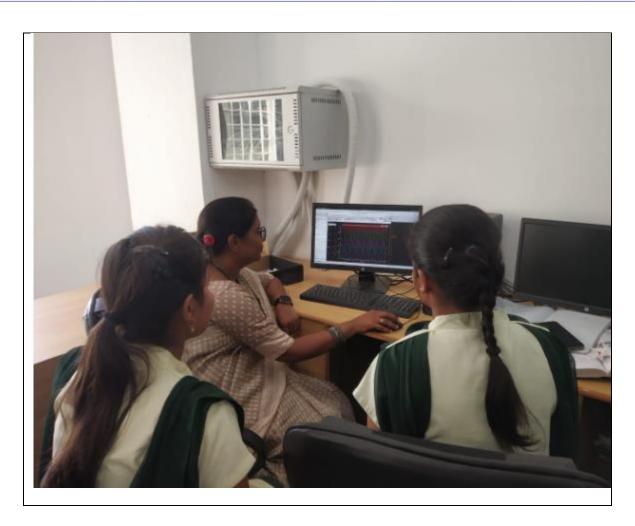








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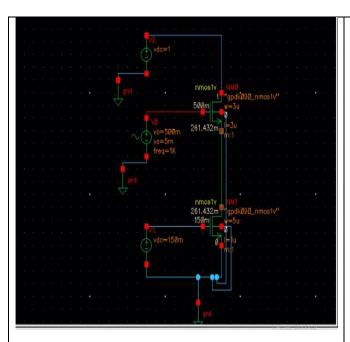


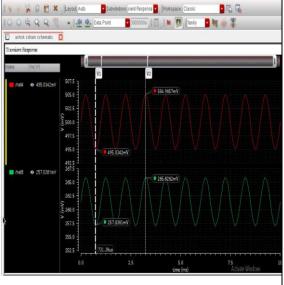


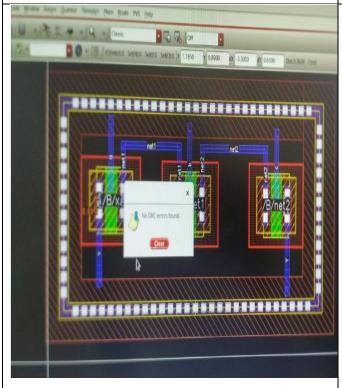


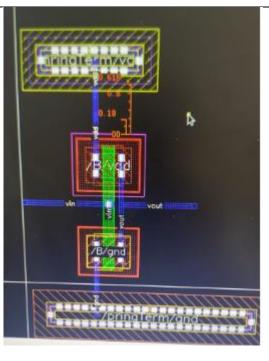


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Centenary Celebrated Sharnbasveshwar Vidya Vardhaka Sangha's

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FACULTY OF ENGINEERING & TECHNOLOGY (CO-ED) DEPARTMENT OF ELECTRONICS & COMMUNICATION ENGINEERING (UNDER IQAC INITIATIVE)

CERTIFICATE

OF ACHIEVEMENT

has successfully completed 12 - weeks value added programme on "INTRODUCTION TO CIRCUIT AND LAYOUT DESIGN" from 28 April 2023 to 09 July 2023.

CHAIRMAN DEPT. ECE SHARNBASVA UNIVERSITY

DEAN OF FACULTY SHARNBASVA UNIVERSITY

DEAN SHARNBASVA UNIVERSITY

REGISTRAR SHARNBASVA UNIVERSITY











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